

Estimating sensitivity of the Einstein@Home search *S5R5*

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Notes on semi-analytically estimating the Hough-on- \mathcal{F} -stat sensitivity of the E@H run *S5R5*.

I. ESTIMATING SENSITIVITY OF HOUGH-ON- \mathcal{F} STATISTIC

We derive an estimate for the sensitivity of the Hough-on- \mathcal{F} statistic [1] used in the E@H search *S5R5*. This builds on the sensitivity-estimation methods developed in [2] and [3], adapted to the **non- χ^2** distributed Hough-on- \mathcal{F} statistic.

The Hough-on- \mathcal{F} statistic is defined as the number $n \leq N$ out of a total of N segments where the \mathcal{F} -statistic value in a template λ crossed a predefined threshold \mathcal{F}_{th} . The phase parameters λ of the templates include the sky-position \vec{n} , frequency f , and higher-order frequency derivatives \dot{f}, \ddot{f}, \dots

The statistic $2\mathcal{F}_k$ in segment k for signals in Gaussian noise follows a χ^2 -distribution with 4 dof and non-centrality parameter ρ_k^2 , i.e.

$$P(2\mathcal{F}|\rho_k) = \chi_4^2(2\mathcal{F}; \rho_k^2), \quad (1)$$

where $\rho_k = \rho_k(\lambda_s, \lambda)$ is the expected SNR in segment k of a signal with phase parameters λ_s when the closest (coarse-grid) template is in λ . This can be expressed in terms of the perfectly-matched SNR $\rho_{\text{opt},k} \equiv \rho_k(\lambda_s, \lambda_s)$ by introducing the per-segment mismatch $\mu_k(\lambda_s, \lambda)$ as

$$\rho_k^2(\lambda_s, \lambda) = [1 - \mu_k(\lambda_s, \lambda)] \rho_{\text{opt},k}^2. \quad (2)$$

The probability of a threshold-crossing in segment k is therefore

$$P(\mathcal{F}_k > \mathcal{F}_{\text{th}}|\rho_k) = \int_{2\mathcal{F}_{\text{th}}}^{\infty} \chi_4^2(2\mathcal{F}; \rho_k^2) d(2\mathcal{F}) = 1 - \text{cdf } \chi_4^2(2\mathcal{F}_{\text{th}}; \rho_k^2). \quad (3)$$

Note that the per-segment SNR ρ_k will not be exactly constant across segments, for several reasons:

- (i) the closest template λ to λ_s will generally be different in every segment, and be subject to a different metric mismatch function μ_k .
- (ii) The optimal SNR $\rho_{\text{opt},k}$ in general varies over segments as a function of their start-time (except if the start-times or segment-lengths are multiples of a sidereal day) due to the time-varying antenna-pattern.
- (iii) in case of non-stationary noise, $\rho_{\text{opt},k}$ further varies over segments as a function of the noise-PSD for that segment S_k .
- (iv) $\rho_{\text{opt},k}$ varies as a function of the amount of data $\Delta T_{\text{data},k}$ used per segment k .

where the antenna-pattern variation (ii) should be very small for the E@H run *S5R5* as $\Delta T = 25$ hours, and we also don't expect large noise-floor variations (iii) in S_k over different segments.

However, in order to be able to continue we have to approximate the per-segment SNR as constant, i.e. $\rho_k \approx \bar{\rho}$, which we write as

$$\bar{\rho}^2 = [1 - \bar{\mu}] \rho_{\text{opt}}^2. \quad (4)$$

We define the per-template per-segment threshold-crossing probability $p_{\bar{\rho}}$ in the presence of a signal with constant per-segment SNR $\bar{\rho}$ as

$$p_{\bar{\rho}}(\mathcal{F}_{\text{th}}; \bar{\rho}) \equiv P(\mathcal{F} > \mathcal{F}_{\text{th}}|\bar{\rho}) = 1 - \text{cdf } \chi_4^2(2\mathcal{F}_{\text{th}}; \bar{\rho}^2), \quad (5)$$

and consequently the per-template per-segment false-alarm probability is $p_0(\mathcal{F}_{\text{th}}) \equiv P(\mathcal{F} > \mathcal{F}_{\text{th}}|\bar{\rho} = 0)$.

Given $p_{\bar{\rho}}$, we can express the (discrete) probability distribution for the Hough number count n as

$$P(n|N, p_{\bar{\rho}}) = \binom{N}{n} p_{\bar{\rho}}^n (1 - p_{\bar{\rho}})^{N-n}. \quad (6)$$

The overall Hough false-alarm and false-dismissal probabilities are therefore

$$p_{f_A}^H(n_{\text{th}}, \mathcal{F}_{\text{th}}) = P(n \geq n_{\text{th}} | \mathcal{F}_{\text{th}}, \bar{\rho} = 0) = \sum_{n=n_{\text{th}}}^N P(n|N, p_0), \quad (7)$$

$$p_{f_D}^H(n_{\text{th}}, \mathcal{F}_{\text{th}}, \bar{\rho}) = P(n < n_{\text{th}} | \mathcal{F}_{\text{th}}, \bar{\rho}) = \sum_{n=0}^{n_{\text{th}}-1} P(n|N, p_{\bar{\rho}}). \quad (8)$$

Estimating the sensitivity of a search typically consists of injecting signals drawn from a population Π_{h_0} with fixed h_0 , and determining the overall false-dismissal probability $p_{f_D}^H$ for this population for a fixed threshold n_{th}^* , corresponding to a certain false-alarm probability $p_{f_A}^{H*}$. The amplitude parameter h_0 is varied until a desired confidence level $1 - p_{f_D}^{H*}$ is obtained, and the corresponding signal amplitude $h_0^* = h_0(p_{f_A}^{H*}, p_{f_D}^{H*})$ is considered a measure for the sensitivity of the search. This critical amplitude is the solution to the equation

$$p_{f_D}^{H*} = P(n < n_{\text{th}}^* | \mathcal{F}_{\text{th}}, h_0^*, \Pi_{h_0}), \quad (9)$$

for given threshold \mathcal{F}_{th} and fixed- h_0 signal population Π_{h_0} .

Following the notation of [2, 4, 5], we can write the optimal (per-segment) SNR ρ_{opt} of a perfectly-matched template ($\lambda = \lambda_s$) as

$$\rho_{\text{opt}} = \hat{\rho}(h_0) \mathcal{R}(\theta), \quad (10)$$

where $\theta \equiv \{\cos \iota, \psi, \vec{n}\}$, and $\mathcal{R}(\theta)$ denotes the geometric antenna-pattern response of the detector network to a GW from direction \vec{n} with amplitude parameters $\{\cos \iota, \psi\}$. Using the notation of Eq. (95) in [5], we can write the response function explicitly as

$$\mathcal{R}^2(\theta) = \frac{25}{4} (\alpha_1 A + \alpha_2 B + 2\alpha_3 C), \quad (11)$$

where $\alpha_i = \alpha_i(\cos \iota, \psi)$ and A, B, C are functions of sky-position \vec{n} (and, generally, data segment k).

For an all-sky search, the signal population Π_{h_0} consists of an isotropic distribution over the sky \vec{n} , and uniform distributions over $\cos \iota \in [-1, 1]$ and $\psi \in [-\pi/4, \pi/4]$, and one can show in general [5] that $\langle \mathcal{R}^2 \rangle_{\theta} = 1$. Following [2] we introduce the optimal per-segment ‘‘rms SNR’’ $\hat{\rho}$ of the signal population Π_{h_0} , namely

$$\hat{\rho} \equiv \sqrt{\langle \rho_{\text{opt}}^2 \rangle_{\theta}} = \frac{2}{5} h_0 \sqrt{\frac{T_{\text{data}}/N}{\mathcal{S}}}, \quad (12)$$

we \mathcal{S} is the overall noise floor, defined as the harmonic mean

$$\mathcal{S}^{-1} \equiv \frac{1}{N_{\text{SFT}}} \sum_{X\alpha}^{N_{\text{SFT}}} S_{X\alpha}^{-1}, \quad (13)$$

over the total number N_{SFT} of SFTs (over all segments), $S_{X\alpha}$ is the per-SFT noise PSD for detector X and time-index α , while

$$T_{\text{data}} \equiv N_{\text{SFT}} T_{\text{SFT}}, \quad (14)$$

is the total amount of data used (over all segments)¹.

The unknown signal location λ_s gives rise to a (template-bank dependent) probability distribution for the mismatch $\bar{\mu}$, which affects the measured (average) per-segment SNR $\bar{\rho}$ as seen in (4). We can absorb this effect by introducing an ‘effective’ response $\mathcal{R}_{\text{eff}}(\theta; \bar{\mu})$ which includes the (unknown) mismatch $\bar{\mu}$:

$$\mathcal{R}_{\text{eff}}(\theta; \bar{\mu}) \equiv [1 - \bar{\mu}(\lambda_s)] \mathcal{R}(\theta), \quad (15)$$

¹ Under ‘‘ideal’’ data conditions of N_{det} detectors with identical stationary noise-floor $S_{X\alpha} = S_n$ without gaps, we have $\mathcal{S} = S_n$, and $T_{\text{data}} = N_{\text{det}} N \Delta T$.

which allows us to express the (per-segment) SNR $\bar{\rho}$ using Eqs. (10) and (4) as

$$\bar{\rho} = \hat{\rho}(h_0) \mathcal{R}_{\text{eff}}(\theta; \bar{\mu}). \quad (16)$$

The sensitivity equation (9) can be written more explicitly as [2]:

$$p_{f_D}^{H*} = \int P(n < n_{\text{th}}^* | \mathcal{F}_{\text{th}}, \bar{\rho} = \hat{\rho}^* \mathcal{R}_{\text{eff}}) P(\mathcal{R}_{\text{eff}} | \Pi_{h_0}) d\mathcal{R}_{\text{eff}}, \quad (17)$$

where the threshold on number-count $n_{\text{th}}^* = n_{\text{th}}(p_{f_A}^{H*}, N)$ is obtained by inverting Eq. (7).

This equation is to be solved for the minimal ‘‘rms-SNR’’ $\hat{\rho}^*$, which we can translate into a minimum signal amplitude h_0^* using Eq. (12):

$$h_0^* = \frac{5}{2} \hat{\rho}^* \sqrt{\frac{\mathcal{S}}{\Delta T_{\text{data}}}}, \quad (18)$$

where $\Delta T_{\text{data}} \equiv T_{\text{data}}/N$.

A. Possible conventions for expressing sensitivity statements

- Karl has proposed [2] to use $\hat{\rho}^*$ directly to characterize the sensitivity of a search.
- Map has used a functional form inspired by the Hough paper (Eq.6.41) in [1]):

$$h_0^* = \frac{F}{N^{1/4}} \sqrt{\frac{\mathcal{S}}{\Delta T_{\text{data}}}}, \quad (19)$$

namely $F \equiv 5/2 \hat{\rho}^* N^{1/4}$.

- We propose to standardize sensitivity statements with respect to only ‘‘global’’ properties of the search, namely \mathcal{S} and T_{data} , and absorb all ‘‘internal’’ search properties (e.g. $N, \Delta T, \bar{\mu}, \dots$) into the sensitivity pre-factor. This could be done either as

$$h_0^* = H \sqrt{\frac{\mathcal{S}}{T_{\text{data}}}}, \quad (20)$$

where $H \equiv 5/2 \hat{\rho}^* \sqrt{N}$.

This is in analogy to what was done for fully-coherent searches (e.g. $H = 11.4$ for a targeted F -statistic search with $p_{f_A}^* = 0.01$ and $p_{f_D}^* = 0.1$ [6]).

- One could even absorb *all* search-specific parameters (including T_{data}) into a single ‘‘sensitivity factor’’ σ_h (with dimensions of $\sqrt{S_n}$, i.e. $\sqrt{\text{Hz}^{-1}}$), and express

$$h_0^* = \frac{\sqrt{\mathcal{S}}}{\sigma_h}, \quad (21)$$

where $\sigma_h^{-1} \equiv 5/2 \hat{\rho}^* \Delta T_{\text{data}}^{-1/2}$. This latter definition has the advantage of quantifying the overall ‘‘intrinsic’’ sensitivity of the search, independently of the detector noise level $\sqrt{\mathcal{S}}$.

B. Biased sensitivity approximation

Note that we can equivalently express Eq. (17) in the form of an average, namely

$$p_{f_D}^{H*} = \langle P(n < n_{\text{th}}^* | \mathcal{F}_{\text{th}}, \bar{\rho} = \hat{\rho}^* \mathcal{R}_{\text{eff}}) \rangle_{\Pi_{h_0}}. \quad (22)$$

A simpler (but biased) sensitivity approximation (used in [1]) proceeds by solving instead

$$p_{f_D}^{H*} = P(n < n_{\text{th}}^* | \mathcal{F}_{\text{th}}, \bar{\rho}^2 = \tilde{\rho}^{*2} \langle \mathcal{R}_{\text{eff}}^2 \rangle_{\Pi_{h_0}}), \quad (23)$$

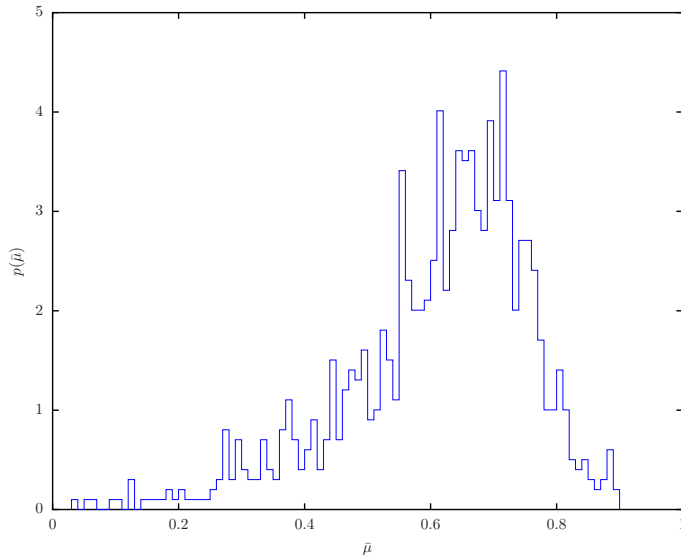


FIG. 1: Effective \mathcal{F} -statistic mismatch distribution in SNR, $P(\bar{\mu}|I)$.

where the averaging refers to the assumed signal population Π_{h_0} . The difference is simply whether to compute the false-dismissal probability for the average SNR², or the average of the false-dismissal probability as a function of SNR². If $P(n < n_{\text{th}}|\bar{\rho})$ (which is always monotonically decreasing with $\bar{\rho}$) is a *convex* function of $\bar{\rho}$ (which won't always be the case), then Jensen's inequality states that

$$P(n < n_{\text{th}}|\hat{\rho}^2 \langle \mathcal{R}_{\text{eff}}^2 \rangle) \leq \langle P(n < n_{\text{th}}|\hat{\rho}^2 \mathcal{R}_{\text{eff}}^2) \rangle, \quad (24)$$

and so one would expect $\tilde{\hat{\rho}}^* \leq \hat{\rho}^*$ for the solutions of (22) and (23), respectively. This corresponds to the biased sensitivity approximation *under-estimating* h_0^* and therefore over-estimating the sensitivity of the search.

Further approximations used [1] to solve Eqs. (7) and (8) for $\bar{\rho}^*$ are: (i) approximate the χ^2 -distributions with Gaussians, and (ii) Taylor-expand in small per-segment SNR $\bar{\rho}$. In [3] we refer to this as the “weak-signal Gaussian” limit. With these approximations, and using $\langle \mathcal{R}_{\text{eff}}^2 \rangle = 1 - \langle \bar{\mu} \rangle$, one can solve Eq. (8) for $\tilde{\hat{\rho}}^* \sqrt{1 - \langle \bar{\mu} \rangle}$, and then use Eq. (18) to obtain the minimal signal amplitude h_0^* .

Apart from the bias introduced by (23), the “weak-signal Gaussian” approximation was found [3] to be rather unreliable for small false-alarm probabilities $p_{f_A}^H$ and segment numbers in the range $N \lesssim 10^3$. We therefore continue with a more exact approach pioneered in [2].

C. Unbiased semi-analytical sensitivity estimation

We can numerically generate the probability distribution $P(\mathcal{R}_{\text{eff}}|\Pi_{h_0})$ for the effective response via Monte-Carlo simulation of the assumed signal- and mismatch distributions. Note that $P(\mathcal{R}|\Pi_{h_0})$ is fully specified in terms of the antenna-pattern response of the detectors for the given signal population. However, we need to prescribe an *ad-hoc* mismatch-distribution $P(\bar{\mu}|I)$ for the Hough-on- \mathcal{F} stat search grids. Drawing values for \mathcal{R}_{eff} is achieved by drawing \mathcal{R} and $\bar{\mu}$ independently, and computing $\mathcal{R}_{\text{eff}} = (1 - \bar{\mu}) \mathcal{R}$.

Given a probability distribution $P(\mathcal{R}_{\text{eff}}|\Pi_{h_0})$ we can numerically solve the integral in Eq. (17) for $\hat{\rho}^*$. Karl has coded this up in octapps, and using this framework we arrive at the following sensitivity estimate.

II. ESTIMATING S5R5 SENSITIVITY

In order to estimate the sensitivity, we use the *S5R5* parameters: $N = 121$, $\mathcal{F}_{\text{th}} = 2.6$. The quoted sensitivity refers to a 90% confidence level, corresponding to $p_{f_D}^{H*} = 0.1$.

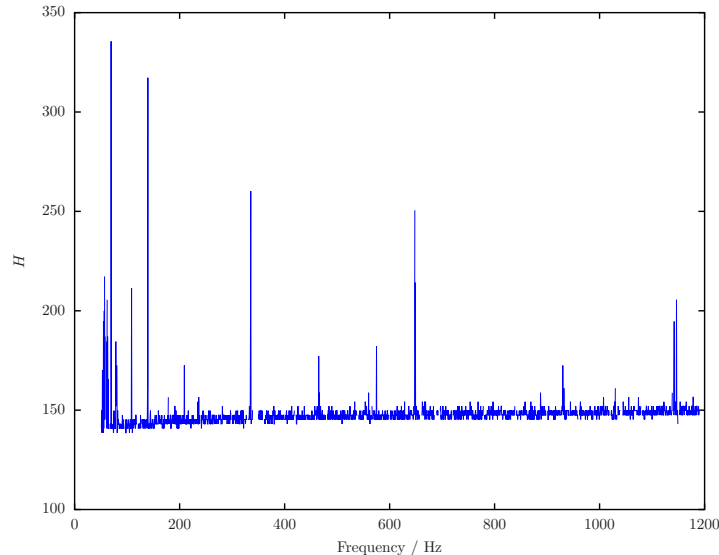


FIG. 2: Predicted sensitivity factor H as a function of frequency.

The two input parameters that need to be specified are the false-alarm probability $p_{f_A}^{H*}$, and the effective \mathcal{F} -statistic mismatch distribution in SNR, $P(\bar{\mu}|I)$:

- The false-alarm probability can equivalently be expressed as a number-count threshold n_{th} , for which we use the values of the loudest candidates found in the *S5R5* post-processing, e.g. see Fig. 3 in the E@H *S5R5* paper. We use $n_{\text{th}} \in [70, 76]$ in the table below, which roughly covers the $\pm 3\sigma$ range in n_{th} .
- We use the mismatch distribution obtained by Miroslav from his follow-up pipeline. The distribution has mean $\langle \bar{\mu} \rangle \approx 0.61$ and standard deviation $\sigma_{\bar{\mu}} \approx 0.15$.

Running `S5R5Sensitivity.m`² which uses `octapps`³, we obtain the following result:

Number-count thresholds	n_{th}	70	71	72	73	74	75	76
False-alarm probabilities	$p_{f_A}^{H*}$	6.0e-13	1.6e-13	3.9e-14	9.4e-15	2.2e-15	5.0e-16	1.1e-16
[Biased (naive) estimate]	$\hat{\rho}_0^*$	2.6	2.7	2.7	2.7	2.7	2.8	2.8
Mean SNR	$\hat{\rho}^*$	5.1	5.2	5.3	5.4	5.4	5.5	5.6
Map's F-factor	F	42.5	43.1	43.8	44.5	45.2	45.8	46.5
Sensitivity factor	H	140.8	143.1	145.3	147.5	149.8	152.0	154.2
Sensitivity scale [$s^{1/2}$]	σ_h	31.0	30.5	30.0	29.6	29.1	28.7	28.3

A. Predicting the *S5R5* upper limits

To predict the h_0 upper limits obtained by the *S5R5* search, we additionally need:

- The number count threshold n_{th} in every 0.5 Hz frequency band for which upper limits are quoted. These were supplied by Paola as significances/critical ratios CR, which were converted to number counts using $n_{\text{th}} = \sigma \text{CR} + \langle n_{\text{th}} \rangle$, with $\sigma = 4.8$ and $\langle n_{\text{th}} \rangle = N(1 + \mathcal{F}_{\text{th}})e^{-\mathcal{F}_{\text{th}}}$. This gave 31 distinct values of n_{th} , ranging from 69 to 120.

² git-version a84e432abebdff6e800861caff59380cec0fde0d-CLEAN

³ git-version 8c592f0ea8084e77f6fbf9e0ffc54ff3d19a9133-CLEAN

- The noise floor \mathcal{S} , harmonically averaged over SFTs and detectors; this was supplied by Map and Paola.

For each 0.5 Hz frequency band, we calculate the sensitivity factor H using the appropriate n_{th} . Then using the mean noise floor \mathcal{S} in the 0.5 Hz band, and Eq. (20), we calculate a prediction for h_0 . The predicted sensitivity factors H are plotted as a function of frequency in Fig. 2.

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